

## 5.0kV<sub>RMS</sub> Opto-Compatible Single Channel Isolated Gate Driver

### **GENERAL DESCRIPTION**

The SLM34xC isolated driver family is an optocompatible, single channel, isolated MOSFET, IGBT gate driver with different drive current capability and UVLO voltage level. The peak output currents are from 1.0A to 4.0A. Key features and characteristics bring significant performance and reliability upgrades over standard opto-coupler based gate drivers while maintaining pin-to-pin compatibility in both schematic and layout design. Performance highlights include high common mode transient immunity (CMTI), low propagation delay, and small pulse width distortion.

The input stage is an emulated diode which means long term reliability and excellent aging characteristics compared to traditional LEDs. It is offered in a SOP6W package with ≥8.0mm creepage and clearance. A mold compound from material group II which has a comparative tracking index (CTI) >400V. High performance and reliability of the SLM34xC makes it ideal for use in all types of motor drives, solar inverters, industrial power supplies, and appliances.

### **APPLICATION**

- AC and brushless DC motor drives
- Renewable energy inverters
- Industrial power supplies

#### **FEATURES**

- 1.0A to 4.0A peak output current
- 110ns (Max.) propagation delay
- 25ns (Max.) part-to-part delay matching
- 35ns (Max.) pulse width distortion
- 150kV/us (Min.) common mode transient immunity (CMTI)
- Wide gate drive supply voltage
  - 10 V to 40 V for SLM345/6C
  - 14 V to 40 V for SLM340/1/3C
- 30V reverse polarity voltage handling capability on input stage
- Pin to pin compatible to opto-coupler isolated gate drivers
- SOP6W package with ≥8.0mm creepage and clearance
- Junction temperature, T<sub>J</sub>: -40°C to +150°C
- Safety certifications
  - 5kVRMs isolation for 1 minute per UL 1577
  - CQC certification per GB4943.1-2022
  - DIN VDE 0884-17: 2021-10

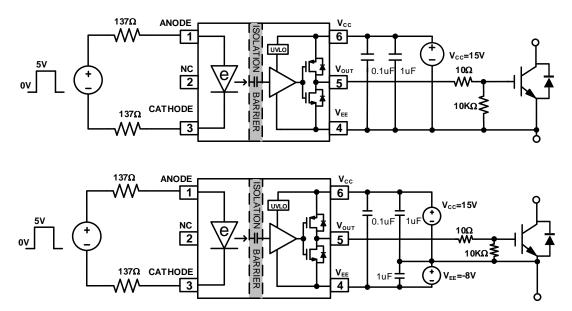


Figure 1. SLM34xC Single and Bipolar Power Supplies Application Circuit to Drive IGBT



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## **PIN CONFIGURATION**

Package	Pin Configuration (Top View)
SOP6W	ANODE 1 0 6 Vcc NC 2 Vout CATHODE 1 VeE

## **PIN DESCRIPTION**

No.	Pin	Description
1	ANODE	Anode
2	NC	No Connection
3	CATHODE	Cathode
4	V <sub>EE</sub>	Negative Power Supply Rail
5	Vouт	Gate Drive Output
6	Vcc	Positive Power Supply Rail

## **FUNCTIONAL BLOCK DIAGRAM**

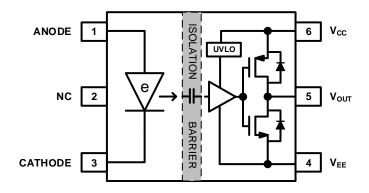


Figure 2. SLM34xC Functional Block Diagram



# **ORDERING INFORMATION**

Order Part No.	Package	QTY
SLM340CCK-DG	SOP6W, Pb-Free	1000/Reel
SLM341CCK-DG	SOP6W, Pb-Free	1000/Reel
SLM343CCK-DG	SOP6W, Pb-Free	1000/Reel
SLM345CCK-DG	SOP6W, Pb-Free	1000/Reel
SLM346CCK-DG	SOP6W, Pb-Free	1000/Reel

# **FAMILY OVERIEW**

Part Number	Peak Output Current	UVLO	Isolation Rating
SLM340C	1.0 A	12.5V	5.0kVrms
SLM341C	3.0 A	12.5V	5.0kVrms
SLM343C	4.0 A	12.5V	5.0kVrms
SLM345C	1.0 A	8.5V	5.0kVrms
SLM346C	4.0 A	8.5V	5.0kVrms



## **ABSOLUTE MAXIMUM RATINGS**

Symbol	Definition	Min	Max	Unit
I <sub>F(AVG)</sub>	Average Input Current		25	mA
V <sub>R</sub>	Reverse Input Voltage		30	V
Vcc -Vee	Output supply voltage		45	V
ΤJ	Junction temperature	-40	150	°C
Ts	Storage temperature	-55	150	

## **RECOMMENDED OPERTION CONDITIONS**

Symbol	Definition	Min	Max	Unit
V <sub>CC</sub> -V <sub>EE</sub>	Output Supply Voltage (SLM340C/SLM341C/ SLM343C)	14 40 \		V
	Output Supply Voltage (SLM345C/SLM346C)	10	40	V
I <sub>F</sub> (ON)	Input Diode Forward Current (Diode "ON")	7	16	mA
V <sub>F</sub> (OFF)	Anode Voltage - Cathode Voltage (Diode "OFF")	-30	0.9	V
TJ	Junction temperature	-40	150	°C
TA	Ambient temperature	-40	125	°C

## **ESD RATINGS**

Symbol	Definition	Value	Unit
V <sub>ESD</sub>	НВМ	±4000	V
V LOD	CDM	±2000	v

## THERMAL INFORMATION

Symbol	Definition	Value	Unit
RөJA	Junction to ambient thermal resistance	125	°C/W
Rejc	Junction to case (top) thermal resistance	66	°C/W
Ψ <sub>ЈТ</sub>	Junction to top characterization parameter	30	°C/W



# **PACKAGE SPECIFICATIONS**

Symbol	Definition	Min	Тур	Max	Units
Rıo	Resistance (Input Side to Output Side)		10 <sup>12</sup>		Ω
Сю	Capacitance (Input Side to Output Side)		0.8		pF
CIN	Input Capacitance		30		pF

## **INSULATION SPECIFICATIONS**

Symbol	Definition	Test Condition	Value	Units
CLR	External clearance	Shortest terminal to terminal distance through air	≥8	mm
CPG	External creepage	Shortest terminal to terminal distance across the package surface	≥8	mm
DTI	Distance through the insulation	Minimum internal gap	>16	um
СТІ	Comparative tracking index	DIN EN 60112 (VDE 0303-11), IEC 60112	>400	V
	Material Group		II	
		Rated mains voltages ≤150Vrms	IV	
		Rated mains voltages ≤300Vrms	IV	
	Overvoltage category	Rated mains voltages ≤600Vrms	III	
		Rated mains voltages ≤1000Vrms	II	
DIN V VDI	E 0884-11 <sup>(1)</sup>			1
Viorm	Maximum repetitive peak isolation voltage		1414	V <sub>PK</sub>
$V_{\text{IOWM}}$	Maximum isolation working voltage		1000	V <sub>RMS</sub>
V <sub>ІОТМ</sub>	Maximum transient isolation voltage	60s	7000	V <sub>PK</sub>
V <sub>IOSM</sub>	Maximum surge isolation voltage	Test method per IEC62368, 1.2/50us waveform, V <sub>TEST</sub> =1.6 x V <sub>IOSM</sub>	6250	V <sub>PK</sub>
<b>q</b> <sub>pd</sub>	Apparent charge	Method b2: $V_{pd(m)}$ =1.875 x $V_{IORM}$ , $tm$ =1 s	≤5	pC
	Climatic Category		40/125/21	
	Pollution Degree		2	
UL1577				
Viso	Withstand Isolation Voltage	V <sub>TEST</sub> =V <sub>ISO</sub> , t=60s (qualification), V <sub>TEST</sub> =1.2 x V <sub>ISO</sub> , t=1s (100% production)	5000	V <sub>RMS</sub>

<sup>1.</sup>Certification pending



## **SAFETY RELATED CERTIFICATIONS**

VDE	UL	CQC
DIN VDE 0884-17: 2021-10	UL 1577 component recognition	Certified according to
	program	GB4943.1-2022
Reinforced Insulation	Single protection, 5000 V <sub>RMS</sub>	Reinforced insulation,
V <sub>IORM</sub> = 1414 V <sub>PK</sub>		Altitude ≤ 5000m, Tropical climate, 400 V <sub>RMS</sub> maximum working voltage
VIOTM = 7000 VPK		400 VRMS Maximum Working Voltage
VIOSM = 6250 VPK		
Pending	Pending	Pending

## **SAFETY LIMITING VALUES**

Symbol	Parameter	Condition	Value	Unit
Is	Safety input, output, or supply	R <sub>θJA</sub> =125°C/W, V <sub>CC</sub> -V <sub>EE</sub> = 15V, T <sub>J</sub> =150°C, T <sub>A</sub> =25°C	50	mA
15	current	R <sub>θJA</sub> =125°C/W, V <sub>CC</sub> -V <sub>EE</sub> = 30V, T <sub>J</sub> =150°C, T <sub>A</sub> =25°C	25	mA
Ps	Safety input, output, or total power	R <sub>0JA</sub> =125°C/W, T <sub>J</sub> =150°C, T <sub>A</sub> =25°C	750	mW
Ts	Maximum safety temperature		150	°C

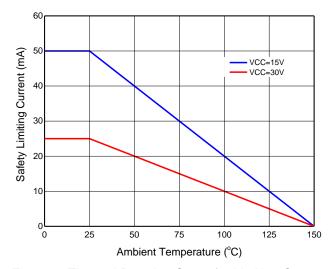


Figure 3. Thermal Derating Curve for Limiting Current per VDE

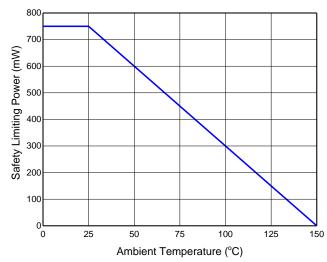


Figure 4. Thermal Derating Curve for Limiting Power per VDE



# **ELECTRIAL CHARACTERISTICS (DC)**

 $V_{CC}$ - $V_{EE}$  = 15V,  $V_{EE}$  = GND and  $T_A$  = 25°C unless otherwise specified. All min and max specifications are at  $T_A$  = -40°C to 125°C

Symbol	Parameter	Condition	Min	Тур	Max	Unit
INPUT				<u> </u>	<u> </u>	
I <sub>FLH</sub>	Input Forward Threshold Current Low to High		1.4	1.9	3	mA
V <sub>F</sub>	Input Forward Voltage	I <sub>F</sub> =10mA	2.1	2.25	2.5	V
ΔV <sub>F</sub> /ΔΤ	Temp Coefficient of Input Forward Voltage	I <sub>F</sub> =10mA		0.5		mV/°C
VR	Input Reverse Breakdown Voltage	I <sub>R</sub> =10uA	30			V
OUTPUT(S	SLM340C/SLM345C)	<u> </u>				
Іон	High Level Peak Output Current	Vcc=15V, I <sub>F</sub> =10mA, C <sub>VDD</sub> =10uF, C <sub>LOAD</sub> =220nF		1.0		А
loL	Low Level Peak Output Current	$V_{CC}$ =15V, $V_{F}$ =0V, $C_{VDD}$ =10uF, $C_{LOAD}$ =220nF		1.5		А
V <sub>OH</sub>	High Level Output Voltage	I <sub>F</sub> =10mA, I <sub>O</sub> =-20mA "With respect to Vcc"		105		mV
Vol	Low Level Output Voltage	V <sub>F</sub> =0V, I <sub>O</sub> =20mA		50		mV
OUTPUT(S	SLM341C)	<u> </u>	<u> </u>		<u> </u>	
Іон	High Level Peak Output Current	$V_{CC}$ =15V, $I_F$ =10mA, $C_{VDD}$ =10uF, $C_{LOAD}$ =220nF		3.0		А
loL	Low Level Peak Output Current	V <sub>CC</sub> =15V, V <sub>F</sub> =0V, C <sub>VDD</sub> =10uF, C <sub>LOAD</sub> =220nF		4.5		А
Vон	High Level Output Voltage	I <sub>F</sub> =10mA, I <sub>O</sub> =-20mA "With respect to Vcc"		33		mV
Vol	Low Level Output Voltage	V <sub>F</sub> =0V, I <sub>O</sub> =20mA		16		mV
OUTPUT(S	SLM343C/SLM346C)		l			1
Іон	High Level Peak Output Current	$V_{CC}$ =15V, $I_F$ =10mA, $C_{VDD}$ =10uF, $C_{LOAD}$ =220nF		4.0		А
loL	Low Level Peak Output Current	V <sub>CC</sub> =15V, V <sub>F</sub> =0V, C <sub>VDD</sub> =10uF, C <sub>LOAD</sub> =220nF		6.0		А
Vон	High Level Output Voltage	I <sub>F</sub> =10mA, I <sub>O</sub> =-20mA "With respect to Vcc"		26		mV
V <sub>OL</sub>	Low Level Output Voltage	V <sub>F</sub> =0V, I <sub>O</sub> =20mA		13		mV



Symbol	Parameter	Condition	Min	Тур	Max	Unit
UNDER VO	DLTAGE LOCKOUT(SLM340C/SLM341C/	SLM343C)				
UVLOR	Under Voltage Lockout Vcc rising	I <sub>F</sub> =10mA	11.5	12.5	13.5	V
UVLO <sub>F</sub>	Under Voltage Lockout Vcc falling	I <sub>F</sub> =10mA	10.5	11.5	12.5	V
UVLO <sub>HYS</sub>	Under Voltage Lockout Hysteresis			1.0		V
UNDER VO	DLTAGE LOCKOUT (SLM345C/SLM346C	)	l			
UVLOR	Under Voltage Lockout Vcc rising	I <sub>F</sub> =10mA	8	8.5	9	V
UVLO <sub>F</sub>	Under Voltage Lockout Vcc falling	I <sub>F</sub> =10mA	7	7.5	8	V
UVLO <sub>HYS</sub>	Under Voltage Lockout Hysteresis			1.0		V

# **SWITCHING CHARACTERISTICS (AC)**

 $V_{\text{CC}}$ - $V_{\text{EE}}$  = 15V,  $V_{\text{EE}}$  = GND and  $T_{\text{A}}$  = 25°C unless otherwise specified. All min and max specifications are at  $T_{\text{A}}$  = -40°C to 125°C

Symbol	Parameter	Condition	Min	Тур	Max	Unit
tplH	Propagation delay, Low to High			75	110	ns
tphL	Propagation delay, High to Low	C <sub>LOAD</sub> =1nF, f <sub>sw</sub> =20kHz,		75	110	ns
tr	Turn on rise time	(50% Duty Cycle),			25	ns
t <sub>f</sub>	Turn off fall time				15	ns
t <sub>PWD</sub>	Pulse Width Distortion				35	ns
t <sub>PDD</sub>	Propagation Delay Difference Between Any Two Parts				25	ns
tuvlo_rec	UVLO Recovery Delay	Vcc Rising from 0V to 15V		22	30	us
CMTI <sub>H</sub>	Output High Level Common Mode Transient Immunity IF=10mA, V <sub>CM</sub> =1000V, V <sub>CC</sub> =15V, T <sub>A</sub> =25°C		150	200		kV/us
CMTI∟	Output Low Level Common Mode Transient Immunity  V <sub>F</sub> =0V, V <sub>CM</sub> =1000V V <sub>CC</sub> =15V, T <sub>A</sub> =25°C		150	200		kV/us



### PARAMETER MEASUREMENT INFORMATION

### Propagation Delay, Rise Time and Fall Time

Figure 5 shows the propagation delay from the input forward current  $I_F$ , to  $V_{OUT}$ . This figure also shows the circuit used to measure the rise ( $t_F$ ) and fall( $t_F$ ) times and the propagation delays  $t_{PLH}$  and  $t_{PHL}$ .

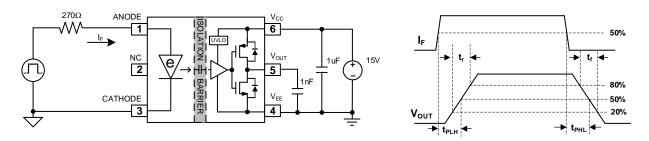


Figure 5. Propagation Delay, Rise Time and Fall Time

### **IOH and IOL Testing**

Figure 6 shows the circuit used to measure the output drive current  $I_{OL}$  and  $I_{OH}$ . A load capacitance of 220nF is used at the output. The peak dv/dt of the capacitor voltage is measured in order to determine the peak source and sink currents of the gate driver.

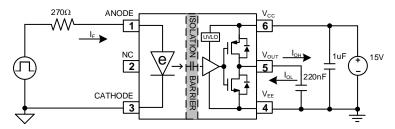


Figure 6. IoH and IoL

### **CMTI Testing**

Figure 7 is the simplified diagram of the CMTI testing. Common mode voltage is set to 1000V. The test is performed with  $I_F=10mA$  ( $V_{OUT}=High$ ) and  $V_F=0V$  ( $V_{OUT}=Low$ ).

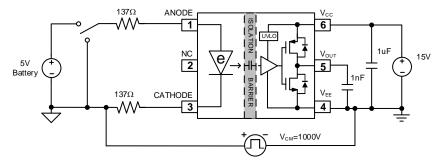


Figure 7. CMTI Test Circuit



### FEATURE DESCRIPTION

SLM34xC is a single channel isolated gate driver, with an opto-compatible input stage, that can drive IGBTs and MOSFETs. It has 1.0 A to 4.0A peak output current capability with maxim output driver supply voltage of 40V. The inputs and the outputs are galvanically isolated. SLM34xC is offered in SOP6W package with >8.5mm creepage and clearance. The reinforced isolation rating is 5kV<sub>RMS</sub> for 60 seconds. It is pin-to-pin compatible with standard opto-coupler isolated gate drivers use an LED as the input stage, SLM34xC uses an emulated diode as the input stage which does not use light emission to transmit signals across the isolation barrier. The input stage is isolated from the driver stage by dual, series HV SiO2 capacitors in full differential configuration that not only provides reinforced isolation but also offers great performance of common mode transient immunity >150kV/us. The e-diode input stage along with capacitive isolation technology gives SLM34xC several performance advantages over standard opto-coupler isolated drivers.

- Since the emulated diode does not use light emission for its operation, the reliability and aging characteristics of SLM34xC are naturally superior to those of standard opto-coupler isolated gate drivers.
- Higher ambient operating temperature range of 125°C, compared to only 105°C for most opto-coupler isolated gate drivers
- The e-diode forward voltage drop has less part-to-part variation and smaller variation across temperature.
   Hence, the operating point of the input stage is more stable and predictable across different parts and operating temperature
- Higher common mode transient immunity than opto-coupler isolated gate drivers
- Smaller propagation delay than opto-coupler isolated gate drivers
- Due to superior process controls achievable in capacitive isolation compared to opto-coupler isolation, there is less part-to-part skew in the propagation delay, making the system design simpler and more robust
- Smaller pulse width distortion than opto-coupler isolated gate drivers

### **Input Stage**

The input stage of SLM34xC is an emulated diode. When the emulated diode is forward biased by applying a positive voltage to the Anode with respect to the Cathode, a forward current,  $I_F$ , flows into the e-diode. The forward voltage drop across the e-diode is 2.25V (typ). An external resistor should be used to limit the forward current. The recommended range for the forward current is 7mA to 16mA. When  $I_F$  exceeds the input forward threshold current  $I_{FLH}$  (2.1mA typ), the  $V_{OUT}$  is driver high. If the  $I_F$  is lower than  $I_{FLH}$ , or the voltage between Anode and Cathode is reverse biased, the  $V_{OUT}$  is driven low.

The reverse breakdown voltage of the e-diode is up to 30V. The large reverse breakdown voltage of the e-diode enables SLM34xC to be operated in interlock architecture as shown in Figure 8. The example shows two gate drivers driving a set of IGBTs. The inputs of the gate drivers are connected as shown in Figure 8 and driven by two buffers that are controlled by the MCU. Interlock architecture prevents both the e-diodes from being "ON" at the same time, preventing shoot through in the IGBTs as shown in Figure 9. It also ensures that if both PWM signals are erroneously stuck high (or low) simultaneously, both gate driver outputs will be driven low.

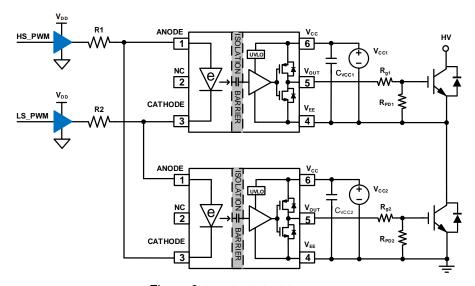
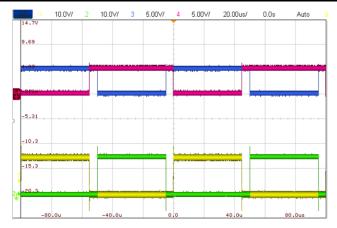


Figure 8. Interlock Architecture





CH1: HS\_OUT, CH2: LS\_OUT, CH3: HS\_PWM, CH4: LS\_PWM

Figure 9. Interlock PWM input and output waveform

### **Under Voltage Lockout (UVLO)**

The SLM34xC integrates the UVLO protection on the  $V_{CC}$  to prevent an under driven condition on IGBTs and MOSFETs. When  $V_{CC}$  is lower than UVLO<sub>R</sub> during start up or lower than UVLO<sub>F</sub> after start up, the UVLO feature holds the  $V_{OUT}$  low, regardless of the input forward current. A hysteresis on the UVLO feature prevents glitch when there is noise from the power supply. When  $V_{CC}$  drops below UVLO<sub>F</sub>, a recovery delay ( $t_{UVLO\_REC}$ ) occurs on the output when the supply voltage rises above UVLO<sub>R</sub> again.

### **Typical Input Configuration Circuit**

The circuit in Figure 10 and Figure 11 show two typical input configuration circuits for SLM34xC to driver IGBT.

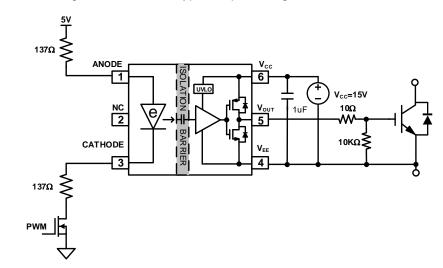


Figure 10. Single MOSFET Circuit as Input Drive of SLM34xC to Drive IGBT

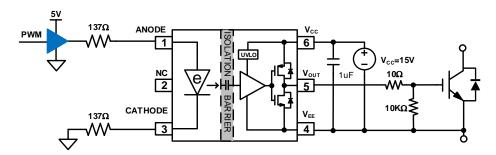


Figure 11. Buffer Circuit as Input Drive of SLM34xC to Drive IGBT



#### Layout

In order to achieve optimum performance for the SLM34xC, some suggestions on PCB layout.

### Component placement:

- Low ESR and low ESL capacitors must be connected close to the device between the V<sub>CC</sub> and V<sub>EE</sub> pins to bypass noise and to support high peak currents when turning on the external power transistor.
- To avoid large negative transients on the VEE pins connected to the switch node, the parasitic inductances between the source of the top transistor and the source of the bottom transistor must be minimized.

### Grounding considerations:

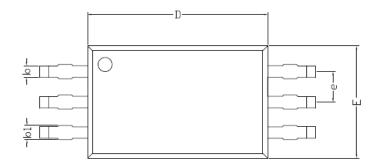
• Limiting the high peak currents that charge and discharge the transistor gates to a minimal physical area is essential. This limitation decreases the loop inductance and minimizes noise on the gate terminals of the transistors. The gate driver must be placed as close as possible to the transistors.

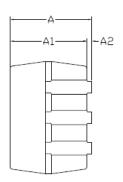
### High-voltage considerations:

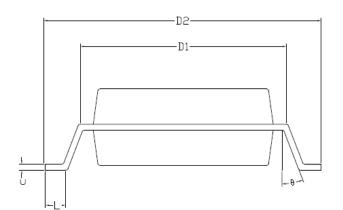
 To ensure isolation performance between the primary and secondary side, avoid placing any PCB traces or copper below the driver device. A PCB cutout or groove is recommended in order to prevent contamination that may compromise the isolation performance.



# **PACKAGE CASE OUTLINES**





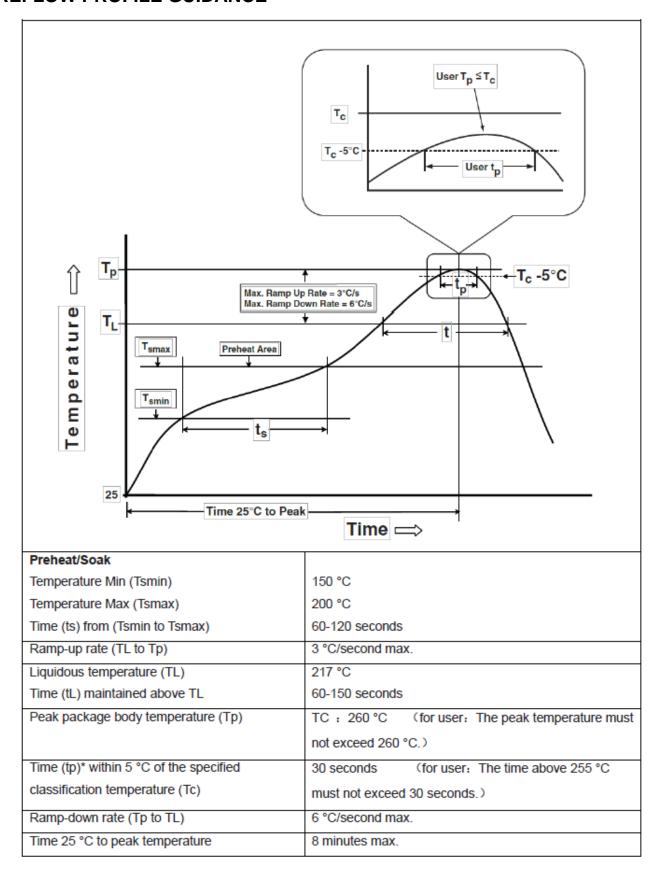


Dimension	MIN	NOM	MAX
A	3. 2	3.4	3.6
A1	3	3. 2	3. 4
A2	0.1	-	0.3
b	0.35	0.4	0.45
b1	0.45	0.48	0.51
С	0.2	0. 25	0.33
D	7.4	7. 5	7.6
D1	8.62	8.72	8.82
D2	11.25	11.5	11.75
Е	4.58	4.68	4.78
е		1.27 BSC	
L	0.5	0.75	1
θ		23° REF	
	Unit : mm		

Figure 12. SOP6W Package Outline Dimensions



## **REFLOW PROFILE GUIDANCE**





## **REVISION HISTORY**

Note: page numbers for previous revisions may differ from page numbers in current version

Page or Item	Subjects (major changes since previous revision)		
Rev 1.0 datasheet: 2024-04-28			
Whole document	Initial datasheet release		