

## Single-Channel, High-Speed, Low-Side Gate Driver

### GENERAL DESCRIPTION

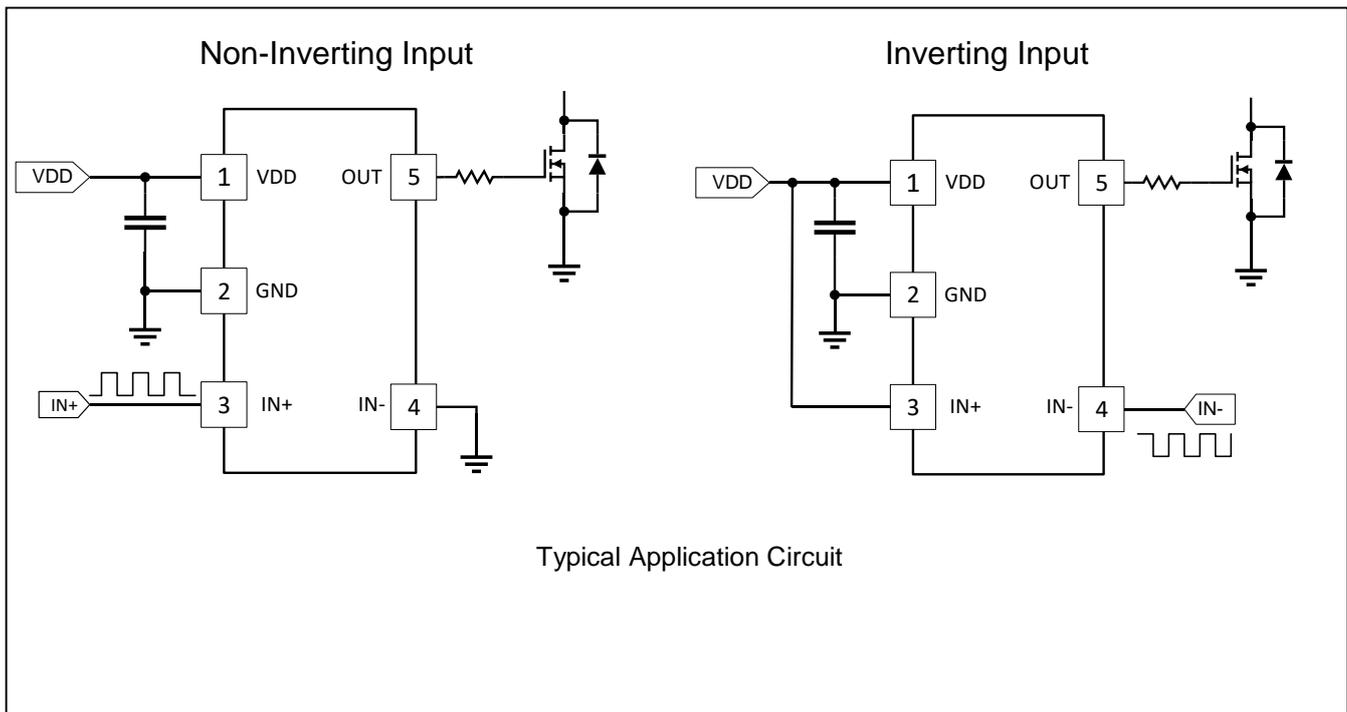
The SiLM27517H-AQ series single-channel, high-speed, low-side gate driver devices can effectively drive MOSFET and IGBT power switches. Using a design that inherently minimizes shoot-through current, SiLM27517H-AQ series products can source and sink high peak-current pulses into capacitive loads offering rail-to-rail drive capability and extremely small propagation delay, typically 18ns.

The SiLM27517H-AQ can provide 4 A source, 5 A sink peak-drive current capability at 15 V VDD supply.

### FEATURES

- Low-cost gate-driver device offering superior replacement of NPN and PNP discrete solutions
- 4 A peak source and 5 A peak sink current
- Fast propagation delay (18 ns typical)
- Fast rise time (9 ns typical)
- Fast fall time (6 ns typical)
- 13.5V to 20V single supply range
- Under-voltage lockout
- TTL and CMOS compatible input logic threshold
- Dual input design (choice of an inverting or non-inverting driver configuration)
- Output held low when input pins are floating
- Operating temperature range of -40°C to 140°C
- SOT23-5, package
- AEC-Q 100 qualified for automotive

### TYPICAL APPLICATION CIRCUIT

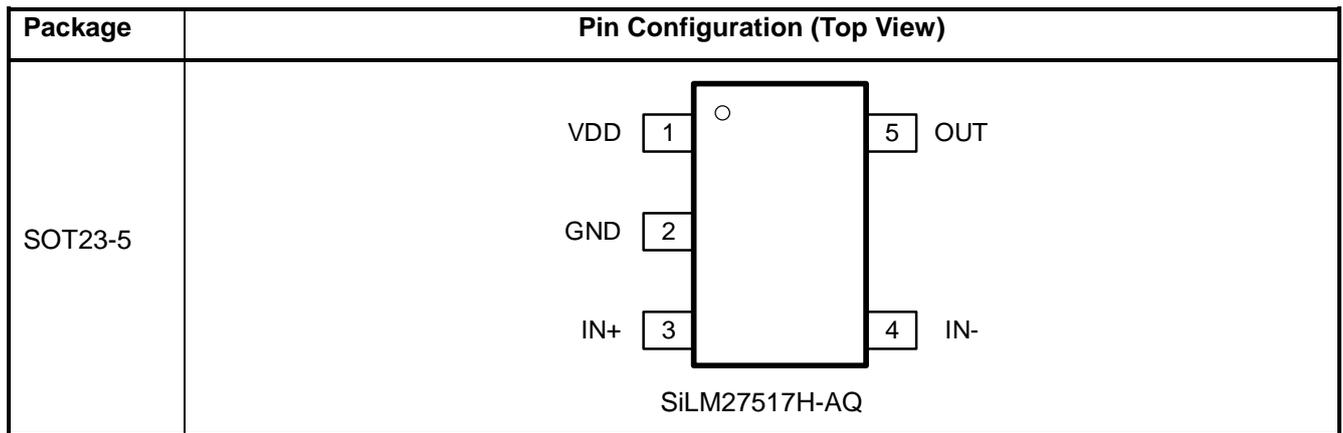


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## PIN CONFIGURATION



## PIN DESCRIPTION

Pin Function		
No.	Name	Function Description
4	IN-	Inverting Input: When the driver is used in noninverting configuration, connect IN- to GND in order to enable output. OUT held LOW if IN- is unbiased or floating.
2	GND	Ground: All signals referenced to this pin.
1	VDD	Bias supply input.
5	OUT	Sourcing/sinking current output of driver
3	IN+	Noninverting Input: When the driver is used in inverting configuration, connect IN+ to VDD in order to enable output. OUT held LOW if IN+ is unbiased or floating.

## ORDERING INFORMATION

Order Part No.	Package	QTY
SiLM27517HAD-AQ	SOT23-5	3000/Reel

**FUNCTIONAL BLOCK DIAGRAM**

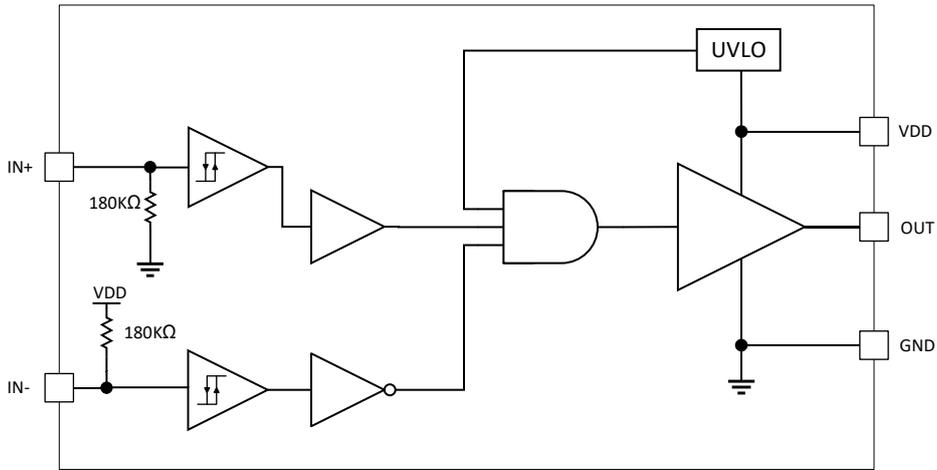


Figure 1. Function Block

## ABSOLUTE MAXIMUM RATINGS<sup>1,2,3</sup>

Over operating free-air temperature range (unless otherwise noted)

Symbol	Description	Min	Max	Unit
V <sub>DD</sub>	Supply Voltage	-0.3	25	V
V <sub>O</sub>	Continuous voltage on OUT	0.3	V <sub>DD</sub> +0.3	
	Repetitive pulse less than 200ns <sup>4</sup>	-2	V <sub>DD</sub> +0.3	
I <sub>O</sub>	Source Continuous Current on OUT		0.3	A
	Source Pulsed Current on OUT (0.5 μs) <sup>4</sup>		-4.5	
	Sink Pulsed Current on OUT (0.5 μs) <sup>4</sup>		5.7	
IN+, IN-	Voltage on the IN+, IN- <sup>5</sup>	-6	25	V
T <sub>J</sub>	Junction temperature	-40	150	°C
T <sub>L</sub>	Lead temperature (soldering, 10 seconds)		300	
T <sub>S</sub>	Storage temperature	-65	150	

- 1) Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- 2) All voltages are with respect to GND unless otherwise noted. Currents are positive into, negative out of the specified terminal.
- 3) These devices are sensitive to electrostatic discharge; follow proper device-handling procedures.
- 4) Values are verified by characterization on bench.
- 5) Maximum voltage on input pins is not restricted by the voltage on the VDD pin.

## RECOMMENDED OPERATION CONDITIONS

Over operating free-air temperature range (unless otherwise noted)

Symbol	Definition	Min	Max	Unit
V <sub>DD</sub>	Supply voltage range	13.5	20	V
IN+, IN-	Input voltage	-5	20	
T <sub>J</sub>	Operation junction temperature range	-40	140	°C

**DYNAMIC ELECTRICAL CHARACTERISTICS**

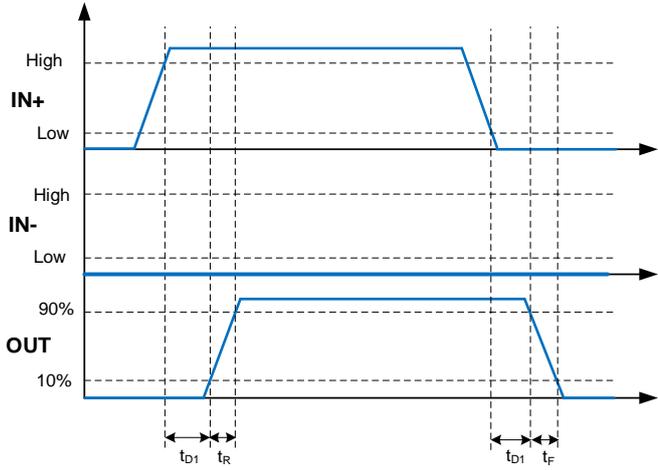
Over operating free-air temperature range (unless otherwise noted)

Symbol	Parameter	Condition	Min.	Typ.	Max.	Unit
$t_{D1}$	IN+ to output propagation delay	$V_{DD}=15V$ , 5V input pulse, $C_{LOAD}=1.8nF$		18	28	ns
$t_{D2}$	IN- to output propagation delay	$V_{DD}=15V$ , 5V input pulse, $C_{LOAD}=1.8nF$		18	28	
$t_R$	Turn-on rise time	$V_{DD}=15V$ , $C_{LOAD}=1.8nF$		9	12	
$t_F$	Turn-off fall time	$V_{DD}=15V$ , $C_{LOAD}=1.8nF$		6	10	

**STATIC ELECTRICAL CHARACTERISTICS**
 $V_{DD}= 15 V$ , 10uF capacitor from VDD to GND.  $T_A = -40^{\circ}C \sim 125^{\circ}C$  unless otherwise specified.

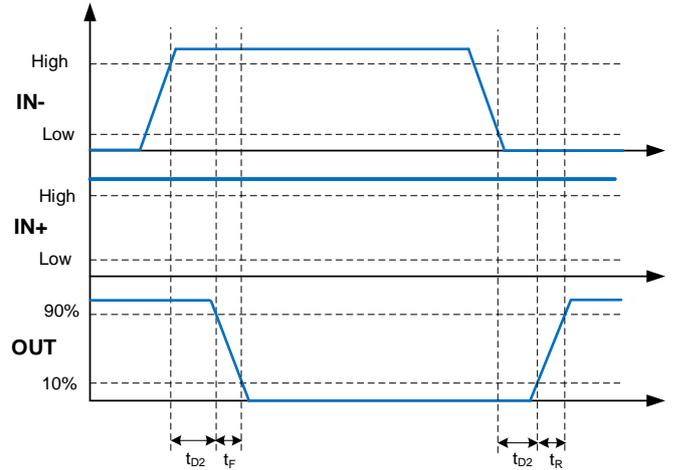
Symbol	Parameter	Condition	Min.	Typ.	Max.	Unit
$V_{IH}$	Logic "1" input voltage	Output high for IN+ pin Output low for IN- pin		2	2.2	V
$V_{IL}$	Logic "0" input voltage	Output low for IN+ pin Output high for IN- pin	1.07	1.3		
$V_{OH}$	High level output voltage, $V_{DD} - V_O$	$V_{DD}=15V$ $I_O = -10$ mA		9	12	mV
$V_{OL}$	Low level output voltage, $V_O$	$V_{DD}=15V$ $I_O = 10$ mA		4.8	7.4	
$I_{DD(off)}$	Startup current	$V_{DD} = 3.4V$ IN+ = $V_{DD}$ , IN- = GND or IN+ = IN- = GND	45	75	120	uA
		IN+ = GND, IN- = $V_{DD}$ or IN+ = IN- = $V_{DD}$	35	60	120	
$V_{DDUV+}$	Undervoltage positive going threshold		11.5	12.5	13.5	V
$V_{DDUV-}$	Undervoltage negative going threshold		10.5	11.5	12.5	
$V_{DD\_H}$	Supply voltage hysteresis			1		
$I_O$	Output high short circuit pulsed current <sup>1</sup>	$V_O = 0 V$ $V_{IN} = \text{Logic "1"}$ $PW \leq 10 \mu s$		-4		A
	Output low short circuit pulsed current <sup>1</sup>	$V_O = 15 V$ $V_{IN} = \text{Logic "0"}$ $PW \leq 10 \mu s$		5		

1) Only bench test.



(PWM input to IN+ pin, IN- pin tied to GND)

Figure 2. Noninverting Configuration



(PWM input to IN- pin, IN+ pin tied to VDD)

Figure 3. Inverting Configuration

## FEATURE DESCRIPTION

### Overview

The SiLM27517H-AQ single-channel, high-speed, low-side gate-driver device can effectively drive MOSFET and IGBT power switches. The device can source and sink high peak-current pulses into capacitive loads, offering rail-to-rail drive capability and extremely small propagation delay of 18 ns (typical). The SiLM27517H-AQ device provides 4A source, 5A sink peak drive current capability. The SiLM27517H-AQ is designed to operate over a wide  $V_{DD}$  range of 13.5V to 20 V, and a wide temperature range of -40°C to 140°C.

The SiLM27517H-AQ device features a dual-input design which offers flexibility of implementing both inverting (IN- pin) and noninverting (IN+ pin) configuration with the same device. Either the IN+ or IN- pin can be used to control the state of the driver output. The unused input pin can be used for enable and disable functions. For system robustness, internal pull up and pull down resistors on the input pins ensure that outputs are held low when the input pins are in floating condition. Therefore, the unused input pin is not left floating and must be properly biased to ensure that driver output is in enabled for normal operation. The input pin threshold of the device is based on TTL and CMOS-compatible low-voltage logic which is fixed and independent of the  $V_{DD}$  supply voltage. Wide hysteresis between the high and low thresholds offers excellent noise immunity.

### VDD and Under-Voltage Lockout

The SiLM27517H-AQ devices have internal UVLO protection feature on the  $V_{DD}$  pin supply circuit blocks. Whenever the driver is in UVLO condition (for example when  $V_{DD}$  voltage is less than  $V_{DDUV+}$  during power up and when  $V_{DD}$  voltage is less than  $V_{DDUV-}$  during power down), this circuit holds all outputs LOW, regardless of the status of the inputs. The UVLO is typically 12.5 V with 1V typical hysteresis. This hysteresis helps prevent chatter when low  $V_{DD}$  supply voltage has noise from the power supply and also when there are droops in the  $V_{DD}$  bias voltage when the system starts switching and there is a sudden increase in  $I_{DD}$ .

For example, at power up, the SiLM27517H-AQ driver output remains LOW until the  $V_{DD}$  voltage reaches the UVLO threshold. The magnitude of the OUT signal rises with  $V_{DD}$  until steady state  $V_{DD}$  is reached. In the noninverting operation (PWM signal applied to IN+ pin), the output remains LOW until the UVLO threshold is reached, and then the output is in phase with the input. In the inverting operation (PWM signal applied to IN- pin), the output remains LOW until the UVLO threshold is reached, and then the output is out phase with the input. In both cases, the unused input pin must be properly biased to enable the output. Note that in these devices the output turns to high state only if IN+ pin is high and IN- pin is low after the UVLO threshold is reached.

Because the driver draws current from the  $V_{DD}$  pin to bias all internal circuits, for the best high speed circuit performance, two  $V_{DD}$  bypass capacitors are recommended to prevent noise problems. The use of surface-mount components is highly recommended. A 0.1 $\mu$ F ceramic capacitor should be located as close as possible to the  $V_{DD}$  to GND pins of the gate driver. In addition, a larger capacitor (such as 1 $\mu$ F or higher value) with relatively low ESR should be connected in parallel and close proximity, in order to help deliver the high-current peaks required by the load. The parallel combination of capacitors characteristic for the expected current levels and switching frequencies in the application.

### Input Stage

The input pins of the SiLM27517H-AQ series products are based on a TTL/CMOS compatible input-threshold logic that is independent of the  $V_{DD}$  supply voltage. With typical high threshold 2V and typical low threshold 1.3V, the logic level thresholds can be conveniently driven with PWM-control signals derived from 3.3V and 5V digital-power controllers. These devices also feature tight control of the input pin threshold voltage levels which eases system-design considerations and ensures stable operation across temperature. The very low input capacitance on these pins reduces loading and increases switching speed.

The device features an important safety function wherein, whenever any of the input pins are in a floating condition, the output of the respective channel is held in the low state. This is achieved using  $V_{DD}$ -pull-up resistors on all the inverting inputs (IN- pin) or GND-pull-down resistors on all the noninverting input pins (IN+ pin).

The device also features a dual-input configuration with two input pins available to control the state of the output. The user has the flexibility to drive the device using either a noninverting input pin (IN+) or an inverting input pin (IN-). The state of the output pin is dependent on the bias on both the IN+ and IN- pins. Once an input pin has been chosen for PWM drive, the other input pin (the unused input pin) must be properly biased in order to enable the output. As mentioned earlier, the unused input pin cannot remain in a floating condition because, whenever any input pin is left in a floating condition, the output is disabled for safety purposes. Alternatively, the unused input pin can effectively be used to implement an enable and disable function, as explained below.

- To drive the device in a noninverting configuration, apply the PWM-control input signal to IN+ pin. In this case, the unused input pin, IN-, must be biased low (for example tied to GND) in order to enable the output. Alternately, the IN- pin can be used to implement the enable/disable function using an external logic signal. OUT is disabled when IN- is biased high and OUT is enabled when IN- is biased low.
- To drive the device in an inverting configuration, apply the PWM-control input signal to IN- pin. In this case, the unused input pin, IN+, must be biased high (for example tied to VDD) in order to enable the output. Alternately, the IN+ pin can be used to implement the enable/disable function using an external logic signal. OUT is disabled when IN+ is biased low and OUT is enabled when IN+ is biased high.
- Finally, note that the output pin is driven into a high state only when IN+ pin is biased high and IN- input is biased low.

The input stage of the driver should preferably be driven by a signal with a short rise or fall time. Caution must be exercised whenever the driver is used with slowly-varying input signals, especially in situations where the device is located in a mechanical socket or PCB layout is not optimal:

- High  $dI/dt$  current from the driver output coupled with board layout parasitic causes ground bounce. Because the device features just one GND pin, which may be referenced to the power ground, the differential voltage between input pins and GND is modified and triggers an unintended change of output state. Because of fast 15ns propagation delay, high-frequency oscillations ultimately occur, which increases power dissipation and poses risk of damage.
- 0.7V input-threshold hysteresis boosts noise immunity compared to most other industry-standard drivers.
- In the worst case, when a slow input signal is used and PCB layout is not optimal, adding a small capacitor(1nF) between input pin and ground very close to the driver device is necessary. This helps to convert the differential mode noise with respect to the input logic circuitry into common-mode noise and avoid unintended change of output state.

If limiting the rise or fall times to the power device is the primary goal, then an external resistance is highly recommended between the output of the driver and the power device instead of adding delays on the input signal. This external resistor has the additional benefit of reducing part of the gate charge related power dissipation in the gate-driver device package and transferring the gate driver into the external resistor.

### Output Stage

The SiLM27517H-AQ device output stage delivers the highest peak-source current when it is most needed during the Miller plateau region of the power-switch turn-on transition (when the power switch drain or collector voltage experiences  $dV/dt$ ). It is capable of supplying 4A peak source and 5A peak sink current pulses. The output voltage swings between VDD and GND providing rail-to-rail operation.

### Power Supply Recommendations

The bias supply voltage ranges for which the SiLM27517H-AQ series products are rated to operate is from 13.5 V to 20 V. The lower end of this range is governed by the internal UVLO protection feature on the  $V_{DD}$  pin supply circuit blocks. Whenever the driver is in UVLO condition when the  $V_{DD}$  pin voltage is below the  $V_{DDUV+}$  supply start threshold, this feature holds the output low, regardless of the status of the inputs. The upper end of this range is driven by the 25 V absolute maximum voltage rating of the  $V_{DD}$  pin of the device (which is a stress rating). Keeping a 5 V margin to allow for transient voltage spikes, the maximum recommended voltage for the  $V_{DD}$  pin is 20 V.

The UVLO protection feature also involves a hysteresis function. This means that when the  $V_{DD}$  pin bias voltage has exceeded the threshold voltage and the device begins to operate, and if the voltage drops, then the device continues to deliver normal functionality unless the voltage drop exceeds the hysteresis specification  $V_{DD\_H(hys)}$ . While operating at or near the 13.5 V range, ensure that the voltage ripple on the auxiliary power supply output is smaller than the hysteresis specification of the device, to avoid triggering a device shutdown.

During system shutdown, the device operation continues until the  $V_{DD}$  pin voltage has dropped below the  $V_{DDUV-}$  threshold, which must be accounted for while evaluating system shutdown timing design requirements. Likewise, at system startup the device does not begin operation until the  $V_{DD}$  pin voltage has exceeded above the  $V_{DDUV+}$  threshold. The quiescent current consumed by the internal circuit blocks of the device is supplied through the  $V_{DD}$  pin. Keep in mind that the charge for source current pulses delivered by the OUT pin is also supplied through the same  $V_{DD}$  pin. As a result, every time a current is sourced out of the output pin (OUT), a corresponding current pulse is delivered into the device through the  $V_{DD}$  pin. Therefore, ensure that local bypass capacitors are provided between the  $V_{DD}$  and GND pins, and located as close to the device as possible for the purpose of decoupling. A

low ESR, ceramic surface mount capacitor is necessary. SiLM27517H-AQ recommends using two capacitors: a 100nF ceramic surface-mount capacitor which can be placed very close to the pins of the device, and another surface-mount capacitor of few microfarads added in parallel.

## Layout Guidelines

Proper PCB layout is extremely important in a high-current fast-switching circuit to provide appropriate device operation and design robustness. The SiLM27517H-AQ gate driver incorporates short-propagation delays and powerful output stages capable of delivering large current peaks with very fast rise and fall times at the gate of the power switch to facilitate voltage transitions very quickly. At higher VDD voltages, the peak current capability is even higher. Very high di/dt causes unacceptable ringing if the trace lengths and impedances are not well controlled. The following circuit layout guidelines are strongly recommended when designing with these high-speed drivers.

- Locate the driver device as close as possible to the power device in order to minimize the length of high-current traces between the output pins and the gate of the power device.
- Locate the VDD bypass capacitors between VDD and GND as close as possible to the driver with minimal trace length to improve the noise filtering. These capacitors support high-peak current being drawn from VDD during turning on the power MOSFET. The use of low inductance SMD components such as chip resistors and chip capacitors is highly recommended.
- The turn-on and turn-off current-loop paths (driver device, power MOSFET and VDD bypass capacitor) should be minimized as much as possible in order to keep the stray inductance to a minimum. High di/dt is established in these loops at two instances, during turn-on and turn-off transients, which will induce significant voltage transients on the output pin of the driver device and gate of the power switch.
- Wherever possible parallel the source and return traces, taking advantage of flux cancellation.
- Separate power traces and signal traces, such as output and input signals.
- Star-point grounding is a good way to minimize noise coupling from one current loop to another. The GND of the driver should be connected to the other circuit nodes such as source of power switch or the ground of PWM controller at one, single point. The connected paths should be as short as possible to reduce inductance and be as wide as possible to reduce resistance.
- Use a ground plane to provide noise shielding. Fast rise and fall times at OUT may corrupt the input signals during transition. The ground plane must not be a conduction path for any current loop. Instead the ground plane must be connected to the star-point with one single trace to establish the ground potential. In addition to noise shielding, the ground plane can help in power dissipation as well.
- In noisy environments, connecting the unused input pin of SiLM27517H-AQ to VDD (in case of IN+) or GND (in case of IN-) using short traces to ensure that the output is enabled and to prevent noise from causing malfunction in the output is necessary.

**PACKAGE CASE OUTLINES**

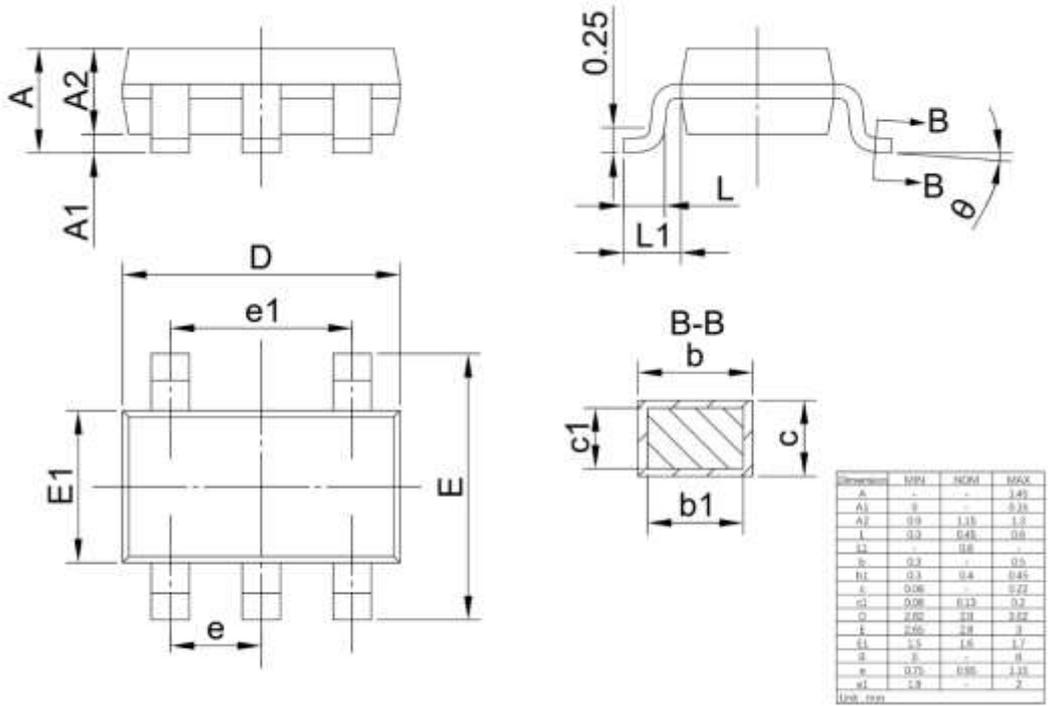


Figure 4. SOT23-5 Outline Dimensions

**REVISION HISTORY**

Note: page numbers for previous revisions may differ from page numbers in current version

<b>Page or Item</b>	<b>Subjects (major changes since previous revision)</b>
<b>Rev 1.0 Datasheet, 2023-09-10</b>	
Whole document	Datasheet initial released